

$$\begin{aligned} \text{BUT } p(x) &= \sum_j p(x|H_j) P(H_j) \\ &= p(x|H_k) P(H_k) \\ &\quad + \sum_{j \neq k} p(x|H_j) P(H_j) \end{aligned}$$

$$\sum_{j \neq k} p(x|H_j) P(H_j) = p(x) - p(x|H_k) P(H_k)$$

↑
↑  
 MINIMIZED                      MAXIMIZED

DECIDE  $H_k$  IF  $p(x|H_k) P(H_k)$   
IS MAXIMIZED FOR  $i=k$  OR

$$p(H_k|x) > p(H_i|x) \quad i \neq k$$

WHAT IS RESULT FOR  $M=2$ ?

SPECIAL CASE:  $P(H_i) = 1/n$

$$P(H_i|x) = \frac{p(x|H_i) P(H_i)}{p(x)} \stackrel{1/n}{\rightarrow}$$

$\Rightarrow$  MAXIMIZE  $p(x|H_i)$  ML RULE

EQUIVALENTLY MAP RULE DECIDES  $H_1$  IF

$$\ln p(\underline{x} | H_1) + \ln P(H_1)$$

IS MAXIMUM.

SEE EXAMPLE 3.6.

### CHAPTER 4 - DETERMINISTIC SIGNALS

$$\begin{array}{ll}
 H_0 : x(n) = w(n) & n = 0, 1, \dots, N-1 \\
 H_1 : x(n) = s(n) + w(n) & n = 0, 1, \dots, N-1
 \end{array}$$

$\uparrow$                        $\uparrow$   
 KNOWN                      WGN  
 DETERMINISTIC  
 SIGNAL

TO MAXIMIZE PD FOR GIVEN PFA  
DECIDE  $H_1$  IF

$$L(\underline{x}) = \frac{p(\underline{x}; H_1)}{p(\underline{x}; H_0)} > \gamma$$

$$p(\underline{x}; H_1) = \frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} (x(n) - s(n))^2}$$

$$p(\underline{x}; H_0) = \frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} x^2(n)}$$

$$L(\underline{x}) = e^{-\frac{1}{2\sigma^2} \left[ -2 \sum_n x(n)s(n) + \sum_n s^2(n) \right]}$$

TAKE LN

$$\frac{1}{\sigma^2} \sum_n x(n)s(n) - \frac{1}{2\sigma^2} \sum_n s^2(n) > \text{LN} \delta$$

↑ NOT A  
FUNCTION  
OF DATA

$$\frac{1}{\sigma^2} \sum_n x(n)s(n) > \text{LN} \delta + \frac{1}{2\sigma^2} \sum_n s^2(n)$$

OR

$$T(\underline{x}) = \sum_n x(n)s(n) > \underbrace{\sigma^2 \text{LN} \delta + \frac{1}{2} \sum_n s^2(n)}_{\gamma'}$$

EXAMPLE : DC LEVEL IN WGN  
 $s(n) = A, A > 0$

$$T(\underline{x}) = A \sum_n x(n) > \gamma'$$

OR

$$T'(\underline{x}) = \frac{1}{NA} T(\underline{x}) = \frac{1}{N} \sum_n x(n) > \gamma''$$

# SAMPLE MEAN

WHAT HAPPENS IF  $A < 0$ ?

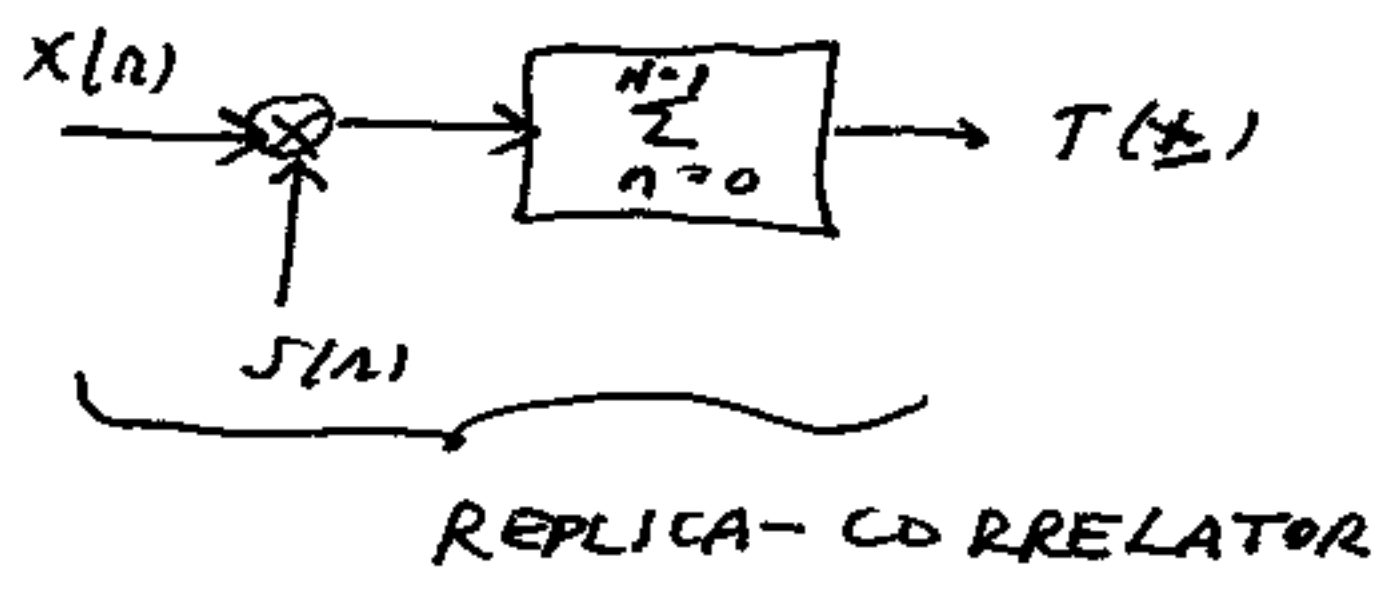
NOTE THAT IN GENERAL WE WEIGHT SAMPLES BY  $w(n)$

$$T(\underline{x}) = \sum_n x(n) w(n)$$

↑ WEIGHTS

WHY?

$$SNR = \frac{S^2(n)}{E(W^2(n))} = \frac{S^2(n)}{\sigma^2}$$



SINCE  $T(\underline{x})$  IS LINEAR FUNCTION OF  $\underline{x} \Rightarrow$  CAN VIEW AS LINEAR FILTER

$$T(x) = \sum_{k=0}^{N-1} s(k) \times x(k)$$

$$y(n) = \sum_{k=0}^n h(n-k) \times x(k)$$

AT  $n = N-1$

$$y(N-1) = \sum_{k=0}^{N-1} h(N-1-k) \times x(k)$$

$\underbrace{\hspace{10em}}_{s(k)}$

LET  $s(k) = h(N-1-k) \quad k = 0, 1, \dots, N-1$

OR  $h(k) = s(N-1-k)$

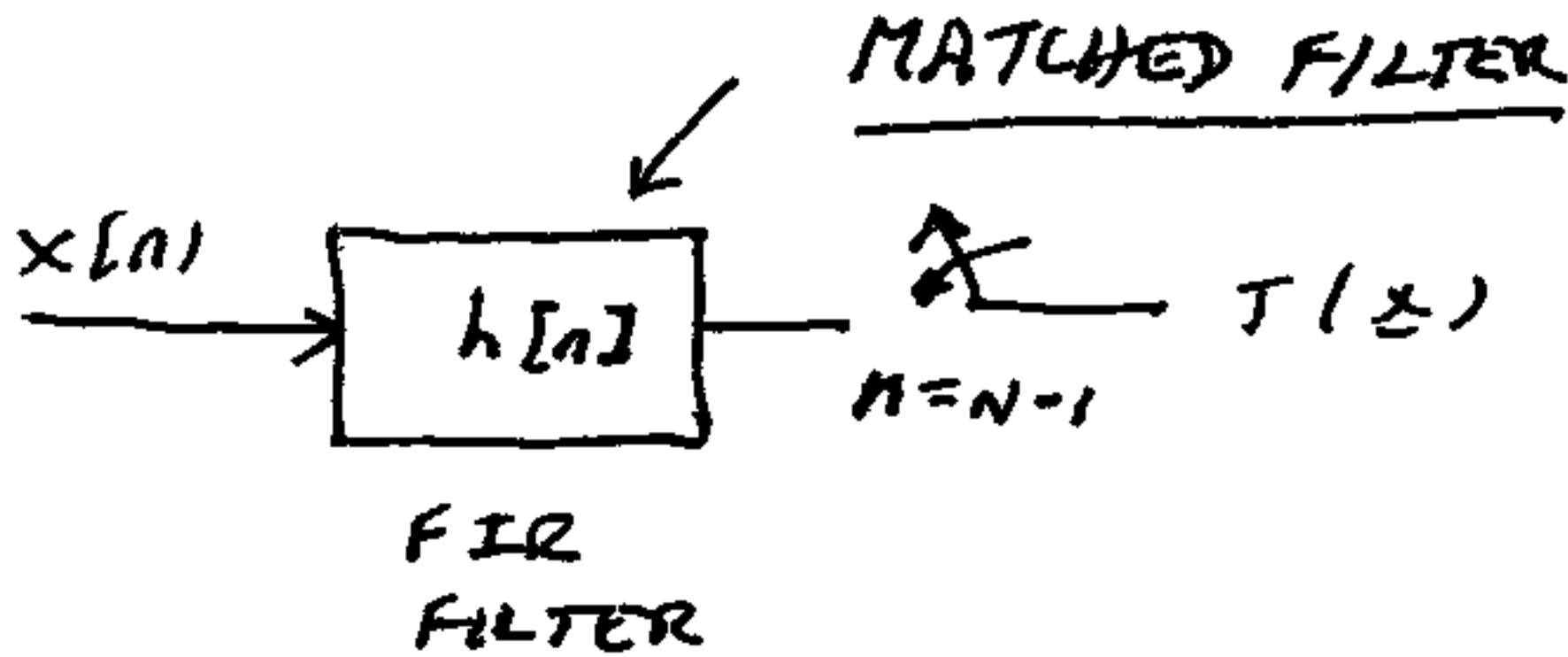
OR  $h(n) = s(N-1-n) \quad n = 0, 1, \dots, N-1$

EXAMPLE :  $N = 5$

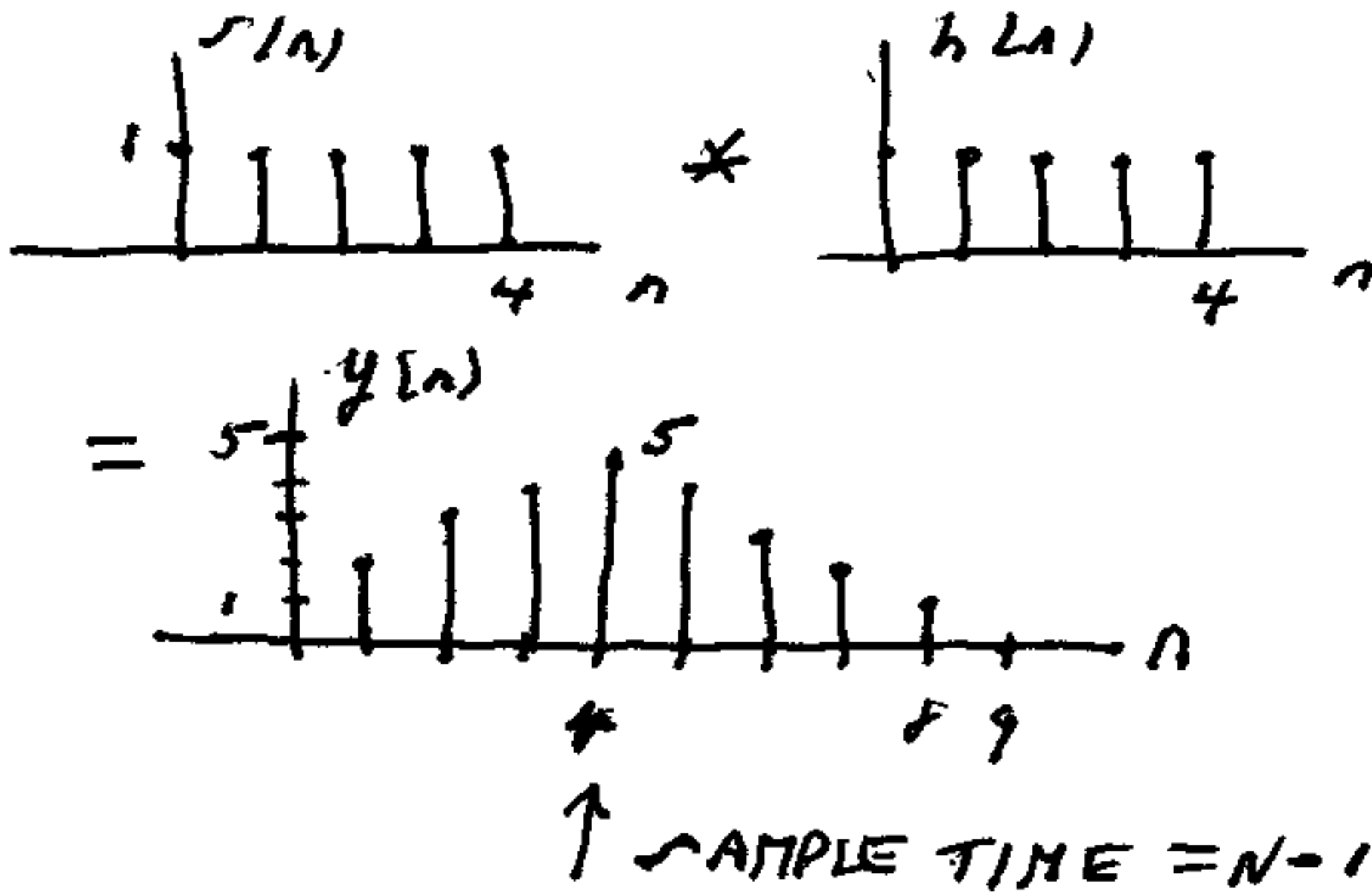


↑ FLIP ABOUT  
 $n=0$  AND SHIFT  
 RIGHT  $(N-1)$  SAMPLES

$h(n)$  IS MATCHED TO  $s(n)$



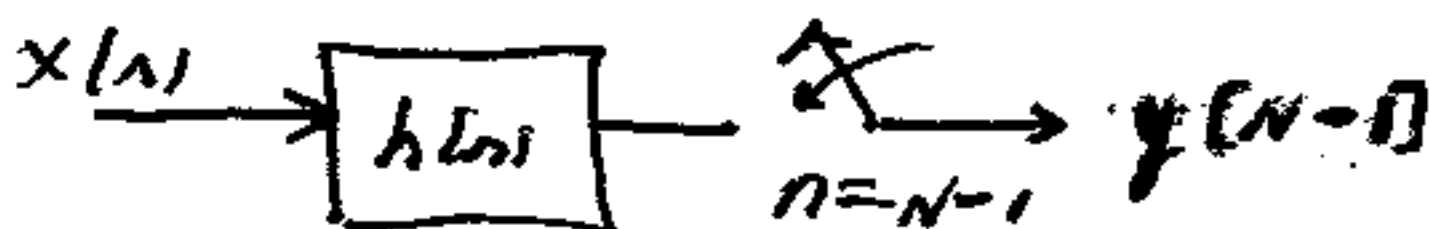
EXAMPLE : DC LEVEL IN WGN  
 $s[n] = 1, N = 5$



IMPORTANT TO SAMPLE AT CORRECT TIME  
 WILL CONSIDER UNKNOWN ARRIVAL TIME LATER!

MAXIMUM SNR PROPERTY

CONSIDER ALL POSSIBLE FIR FILTER  
 DETECTORS OR



$h(n)$  NONZERO OVER  $[0, N-1]$ .  
WHICH  $h(n)$  MAXIMIZES SNR AT  
ITS OUTPUT?

$$\text{SNR} = \frac{E^2(y[N-1])}{\text{VAR}(y[N-1])}$$

DO NOT ASSUME  $h(n) = \delta(N-1-n)$ .

TO MAXIMIZE SNR LET  $\underline{h} = [h[N-1], \dots, h[0]]^T$

$$\underline{s} = [s[0], \dots, s[N-1]]^T$$

$$\text{SNR} = \frac{\left( \sum_{k=0}^{N-1} h[N-1-k] s[k] \right)^2}{E \left[ \left( \sum_{k=0}^{N-1} h[N-1-k] w[k] \right)^2 \right]} = \frac{\text{SIGNAL POWER}}{\text{NOISE POWER}}$$

AT FILTER OUTPUT

$$\begin{aligned} \text{SNR} &= \frac{(\underline{h}^T \underline{s})^2}{E[(\underline{h}^T \underline{w})^2]} = \frac{(\underline{h}^T \underline{s})^2}{E[\underline{h}^T \underline{w} \underline{w}^T \underline{h}]} \\ &= \frac{(\underline{h}^T \underline{s})^2}{\underline{h}^T E(\underline{w} \underline{w}^T) \underline{h}} = \frac{(\underline{h}^T \underline{s})^2}{\sigma^2 \underline{h}^T \underline{h}} \end{aligned}$$

BUT BY CAUCHY-SCHWARZ INEQUALITY

$$(\underline{h}^T \underline{s})^2 \leq (\underline{h}^T \underline{h})(\underline{s}^T \underline{s})$$

WITH = IF AND ONLY IF  $\underline{h} = c \underline{s}$  FOR  
 $c$  A CONSTANT.

$$\Rightarrow \text{SNR} \leq \frac{1}{\sigma^2} \underline{s}^T \underline{s} = \frac{E}{\sigma^2} \quad \left\{ \begin{array}{l} \text{SIGNAL} \\ \text{ENERGY} \end{array} \right.$$

FOR = OR MAXIMUM SNR

$$\underline{h} = c \underline{s} \quad \text{OR}$$

$$h[n] = c \sqrt{N-1-n} \quad n=0, 1, \dots, N-1$$

THUS, MATCHED FILTER MAXIMIZES  
 SNR AT OUTPUT OF ALL FIR FILTERS.

WHAT IF NOISE IS NONGAUSSIAN?

### PERFORMANCE OF MATCHED FILTER

$$\text{DECIDE } H_1 \text{ IF } T(\underline{x}) = \sum_{n=0}^{N-1} x[n] s[n] > \gamma'$$

WISH TO DETERMINE  $\gamma'$  AND  
 PERFORMANCE

$T(\underline{x})$  IS GAUSSIAN SINCE  $x[n]$  IS GAUSSIAN



$$E(T; H_0) = E\left(\sum_n W(n)S(n)\right) = 0$$

$$\begin{aligned} E(T; H_1) &= E\left(\sum_n (S(n) + W(n))S(n)\right) \\ &= \varepsilon \end{aligned}$$

$$\text{VAR}(T; H_0) = \text{VAR}(T; H_1)$$

$$= \text{VAR}\left(\sum_n W(n)S(n)\right)$$

$$= \sum_n S^2(n) \text{VAR}(W(n))$$

$$= \sigma^2 \sum S^2(n) = \sigma^2 \varepsilon$$

$$\Rightarrow \begin{array}{ll} T \sim N(0, \sigma^2 \varepsilon) & H_0 \\ T \sim N(\varepsilon, \sigma^2 \varepsilon) & H_1 \end{array}$$

$$P_{FA} = P_r \{T > \gamma'; H_0\}$$

$$= Q\left(\frac{\gamma'}{\sqrt{\sigma^2 \varepsilon}}\right)$$

$$P_D = P_r \{T > \gamma'; H_1\}$$

$$= Q\left(\frac{\gamma' - \varepsilon}{\sqrt{\sigma^2 \varepsilon}}\right)$$

TO SET PFA CHOOSE

$$\delta' = \sqrt{\sigma^2 \epsilon} Q^{-1}(P_{FA})$$

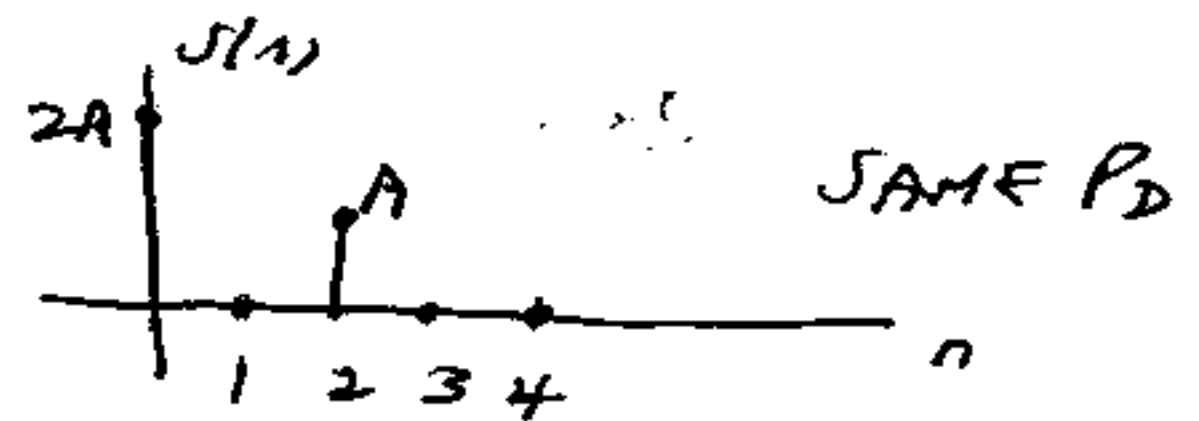
TO FIND PERFORMANCE

$$P_D = Q\left(\frac{\delta'}{\sqrt{\sigma^2 \epsilon}} - \sqrt{\epsilon/\sigma^2}\right)$$

$$= Q\left(Q^{-1}(P_{FA}) - \sqrt{\epsilon/\sigma^2}\right)$$

$\eta = \text{ENERGY-TO-NOISE RATIO}$

PERFORMANCE ONLY DEPENDS ON  
SIGNAL ENERGY NOT SHAPE



ALSO NOTE THAT  $\epsilon/\sigma^2$  IS SNR AT  
MATCHED FILTER OUTPUT.

SHOULD SHOW THAT  $\eta = d^2$  (DEFLECTION  
COEFF.)

## GENERALIZED MATCHED FILTERS

NOW ASSUME NOISE IS GAUSSIAN  
BUT NOT WHITE.

$$\Rightarrow \underline{C} \neq \sigma^2 \underline{I} \quad \underline{C} = \text{COVARIANCE MATRIX OF } W(n)$$

EXAMPLE :  $N = 3$

$$\underline{C} = \begin{bmatrix} E(W^2[0]) & E(W[0]W[1]) & E(W[0]W[2]) \\ E(W[1]W[0]) & E(W^2[1]) & E(W[1]W[2]) \\ E(W[2]W[0]) & E(W[2]W[1]) & E(W^2[2]) \end{bmatrix}$$

FOR WHITE NOISE

$$E(W[m]W[n]) = \begin{cases} \sigma^2 & m=n \\ 0 & m \neq n \end{cases}$$

$$\Rightarrow \underline{C} = \sigma^2 \underline{I}$$

EXAMPLE :  $N = 3$  COLORED NOISE (WSS)

$$E(W[m]W[n]) = \Gamma_w(m-n) = a^{|m-n|}$$

↑  
AUTOCORRELATION

$$\underline{\Sigma} = \begin{bmatrix} 1 & a & a^2 \\ a & 1 & a \\ a^2 & a & 1 \end{bmatrix} \quad |a| < 1$$

NOW FOR ANY  $\underline{\Sigma}$  (MUST BE SYMMETRIC AND POSITIVE DEFINITE)

$$\begin{aligned} \underline{\Sigma} &= E \left[ (\underline{W} - E(\underline{W})) (\underline{W} - E(\underline{W}))^T \right] \\ &\quad \uparrow \qquad \qquad \uparrow \\ &\quad = 0 \qquad \qquad = 0 \\ &= E(\underline{W}\underline{W}^T) \end{aligned}$$

DETERMINE N-P DETECTOR FOR

$$H_0: x(n) = w(n)$$

$$H_1: x(n) = s(n) + w(n)$$

↑  
KNOWN

$$\underline{W} = [w(0) w(1) \dots w(N-1)]^T \quad N \times 1$$

$$\sim N(\underline{0}, \underline{\Sigma})$$

↑                    ↑  
 MEAN                COVARIANCE  
 VECTOR              MATRIX  
 N × 1                N × N

UNDER  $H_0$   $\underline{x} \sim N(\underline{0}, \underline{C})$

UNDER  $H_1$   $\underline{x} \sim N(\underline{s}, \underline{C})$

DECIDE  $H_1$  IF

$$L(\underline{x}) = \ln \frac{p(\underline{x}; H_1)}{p(\underline{x}; H_0)} > \ln \gamma.$$

$$L(\underline{x}) = \ln \frac{\frac{1}{(2\pi)^{N/2} \text{DET}^{1/2}(\underline{C})} e^{-\frac{1}{2}(\underline{x}-\underline{s})^T \underline{C}^{-1}(\underline{x}-\underline{s})}}{\frac{1}{(2\pi)^{N/2} \text{DET}^{1/2}(\underline{C})} e^{-\frac{1}{2}\underline{x}^T \underline{C}^{-1}\underline{x}}}$$

$$= \underline{x}^T \underline{C}^{-1} \underline{s} - \frac{1}{2} \underline{s}^T \underline{C}^{-1} \underline{s}$$

DECIDE  $H_1$  IF

$$T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{s} > \underbrace{\ln \gamma + \frac{1}{2} \underline{s}^T \underline{C}^{-1} \underline{s}}_{\gamma'}$$

EXAMPLE: WGN  $\underline{C} = \sigma^2 \underline{I}$

$$\Rightarrow T(\underline{x}) = \underline{x}^T \frac{1}{\sigma^2} \underline{I} \underline{s}$$

$$= \frac{1}{\sigma^2} \sum_{n=0}^{N-1} x(n) s(n)$$

DECIDE  $H_1$   $H_0$

$$\sum_n x(n) s(n) > \sigma^2 \gamma'$$

REPLICA-CORRELATOR

EXAMPLE:  $w(n) \sim N(0, \sigma_n^2)$

UNCORRELATED WITH UNEQUAL  
VARIANCES

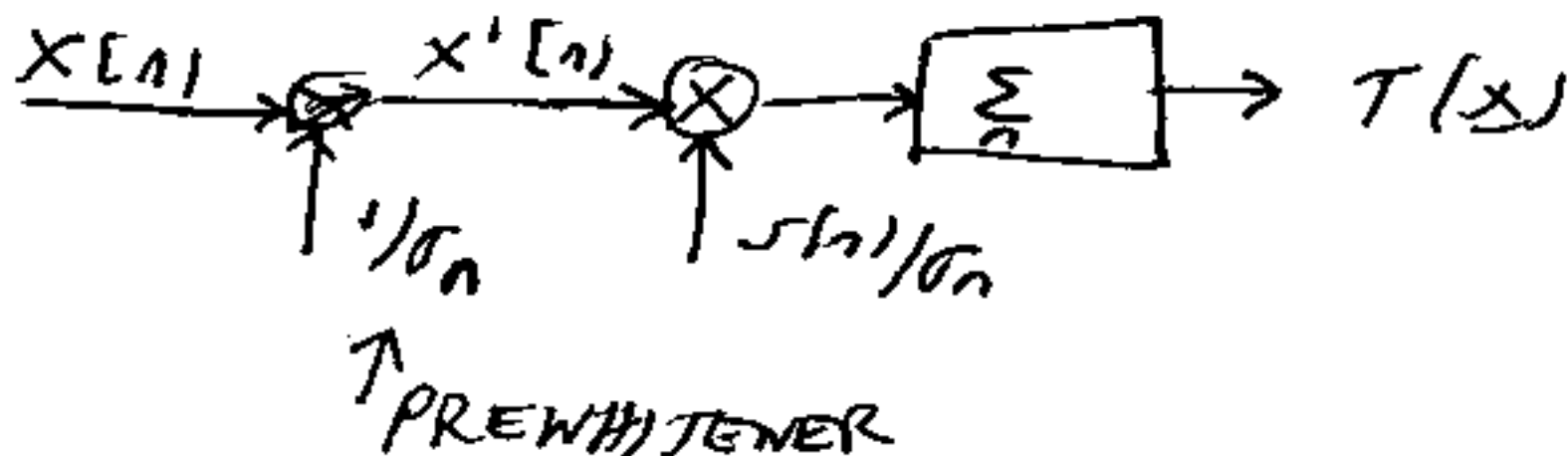
$$\underline{C} = \text{DIAG}(\sigma_0^2, \sigma_1^2, \dots, \sigma_{N-1}^2)$$

$$\underline{C}^{-1} = \text{DIAG}(1/\sigma_0^2, 1/\sigma_1^2, \dots, 1/\sigma_{N-1}^2)$$

$$T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{s}$$

$$= \sum_{n=0}^{N-1} \frac{x(n) s(n)}{\sigma_n^2} \leftarrow \text{WEIGHTS}$$

STATISTIC EMPHASIZES  $x(n) s(n)$   
FOR WHICH  $\sigma_n^2$  IS SMALL.



OPTIMAL DETECTOR FIRST PREWHITENS NOISE SAMPLES

$$\begin{aligned}
 x'[n] &= x[n] / \sigma_n \\
 &= w[n] / \sigma_n \quad H_0 \\
 &= \frac{w[n]}{\sigma_n} + \frac{s[n]}{\sigma_n} \quad H_1
 \end{aligned}$$

BUT  $\frac{w[0]}{\sigma_0}, \frac{w[1]}{\sigma_1}, \dots, \frac{w[N-1]}{\sigma_{N-1}}$  IS WGN.

THEN  $s'[n] = s[n] / \sigma_n$  IS USED AS THE REPLICA SINCE ORIGINAL SIGNAL IS ALSO PREWHITENED.

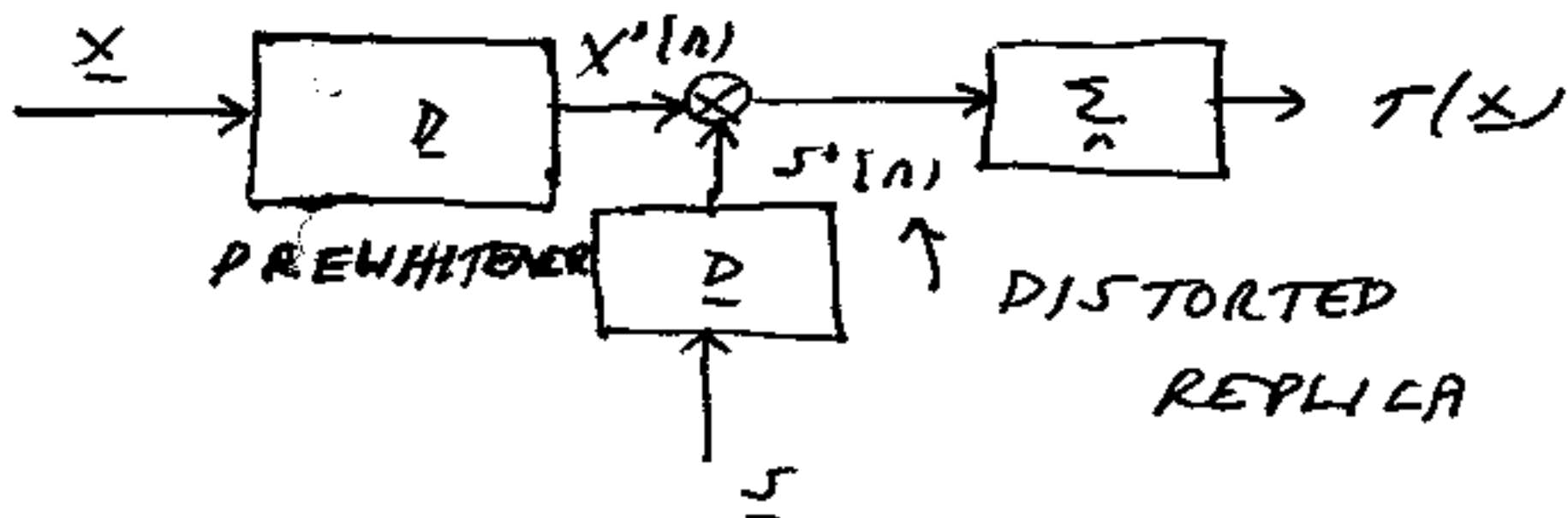
$$T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{s} \quad \text{CALLED A}$$

GENERALIZED M.F. OR CORRELATOR

ALTERNATIVE STRUCTURE :

$$\text{LET } \underline{C}^{-1} = \underline{D}^T \underline{D} \quad \underline{D} \text{ IS NONSINGULAR} \\
 \text{(CHOLESKY FOR EXAMPLE)}$$

$$\begin{aligned}
 T(\underline{x}) &= \underline{x}^T \underline{C}^{-1} \underline{s} \\
 &= \underline{x}^T \underline{D}^T \underline{D} \underline{s} \\
 &= \underline{x}'^T \underline{s}'
 \end{aligned}$$



NOISE AT OUTPUT OF  $\underline{D}$  IS  $\underline{W}$  SINCE

$$\begin{aligned}
 \underline{W}' &= \underline{D} \underline{W} & \underline{C}_{W'} &= E(\underline{W}' \underline{W}'^T) \\
 & & &= E(\underline{D} \underline{W} \underline{W}^T \underline{D}^T) \\
 & & &= \underline{D} \underline{C} \underline{D}^T \\
 & & &= \underline{D} (\underline{D}^T \underline{D})^{-1} \underline{D}^T = \underline{I}
 \end{aligned}$$

### PERFORMANCE

$$T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{s} > \gamma'$$

$T(\underline{x})$  IS GAUSSIAN SINCE

$$\begin{aligned}
 \underline{x} &\sim N(\underline{0}, \underline{C}) & \mathcal{H}_0 \\
 &\sim N(\underline{s}, \underline{C}) & \mathcal{H}_1
 \end{aligned}$$



$$\text{AND } T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{v} = (\underline{v}^T \underline{C}^{-1})^T \underline{x}$$

$$= \underline{a}^T \underline{x} \quad \text{LINEAR TRANSFORMATION}$$

$$E(T; H_0) = E(\underline{w}^T \underline{C}^{-1} \underline{v}) = 0$$

$$E(T; H_1) = E((\underline{v} + \underline{w})^T \underline{C}^{-1} \underline{v}) = \underline{v}^T \underline{C}^{-1} \underline{v}$$

$$\begin{aligned} \text{VAR}(T; H_0) &= E(T^2) \\ &= E((\underline{w}^T \underline{C}^{-1} \underline{v})^2) \\ &= E(\underline{v}^T \underline{C}^{-1} \underline{w} \underline{w}^T \underline{C}^{-1} \underline{v}) \\ &= \underline{v}^T \underline{C}^{-1} E(\underline{w} \underline{w}^T) \underline{C}^{-1} \underline{v} \\ &= \underline{v}^T \underline{C}^{-1} \underline{v} \end{aligned}$$

$$\text{SIMILARLY } \text{VAR}(T; H_1) = \underline{v}^T \underline{C}^{-1} \underline{v}$$

$$T(\underline{x}) \sim N(0, \underline{v}^T \underline{C}^{-1} \underline{v}) \quad H_0$$

$$N(\underline{v}^T \underline{C}^{-1} \underline{v}, \underline{v}^T \underline{C}^{-1} \underline{v}) \quad H_1$$

(GAUSS-BAUSS PROBLEM)

AS BEFORE

$$P_{FA} = Q\left(\frac{\gamma'}{\sqrt{\underline{v}^T \underline{C}^{-1} \underline{v}}}\right)$$

$$P_D = Q\left(\frac{\gamma' - \underline{v}^T \underline{C}^{-1} \underline{v}}{\sqrt{\underline{v}^T \underline{C}^{-1} \underline{v}}}\right)$$

$$\Rightarrow P_D = Q(Q^{-1}(P_{FA}) - \sqrt{d^2})$$

$$d^2 = \underline{s}^T \underline{C}^{-1} \underline{s}$$

EXAMPLE : WGN  $\underline{C} = \sigma^2 \underline{I}$

$$d^2 = \frac{1}{\sigma^2} \underline{s}^T \underline{s} = \epsilon / \sigma^2 = 1$$

PERFORMANCE IMPROVES WITH  $\underline{s}^T \underline{C}^{-1} \underline{s}$   
(NOT SOLELY DEPENDENT ON  $\epsilon = \underline{s}^T \underline{s}$ )

EXAMPLE : UNCORRELATED NOISE  
WITH UNEQUAL VARIANCES

$$\underline{C} = \text{DIAG}(\sigma_0^2, \sigma_1^2, \dots, \sigma_{N-1}^2)$$

$$\underline{s}^T \underline{C}^{-1} \underline{s} = \sum_{n=0}^{N-1} \frac{s^2(n)}{\sigma_n^2}$$

FOR GIVEN  $\sigma_n^2$ 'S HOW CAN WE  
CHOOSE SIGNAL TO MAXIMIZE  
 $\underline{s}^T \underline{C}^{-1} \underline{s}$  (SIGNAL DESIGN PROBLEM)?

MUST CONSTRAIN  $\epsilon = \underline{s}^T \underline{s}$  OR ELSE  
SOLUTION IS TO LET  $s(n) \rightarrow \infty$ .

MAXIMIZE  $\underline{S}^T \underline{C}^{-1} \underline{S}$  WITH  $\underline{S}^T \underline{S} = \epsilon$

SOLUTION: USE LAGRANGIAN MULTIPLIERS

$$F = \sum_n \frac{s^2(n)}{\sigma_n^2} + \lambda (\epsilon - \sum_n s^2(n))$$

$$\frac{\partial F}{\partial s(k)} = \frac{2s(k)}{\sigma_k^2} - 2\lambda s(k) = 0$$

$$\Rightarrow 2s(k) \left( \frac{1}{\sigma_k^2} - \lambda \right)$$

ASSUMING  $s(k) \neq 0 \Rightarrow \lambda = 1/\sigma_k^2$

IF ALL  $\sigma_n^2$ 'S ARE DISTINCT  $\lambda = 1/\sigma_k^2$  CAN HOLD FOR ONLY ONE  $k$ . FOR THE OTHERS  $s(k) = 0$ .

$$\Rightarrow \begin{array}{ll} s(n) = 0 & n \neq k \\ s(k) & n = k \end{array}$$

$$\underline{S}^T \underline{C}^{-1} \underline{S} = \sum \frac{s^2(n)}{\sigma_n^2} = \frac{s^2(k)}{\sigma_k^2}$$

TO MAXIMIZE CHOOSE  $k$  FOR WHICH  $\sigma_k^2$  IS MINIMUM.

$$\text{THEN } S(n) = \begin{cases} \sqrt{E} & n = k \\ 0 & n \neq k \end{cases}$$

CONCENTRATE ALL ENERGY AT SAMPLE WITH SMALLEST AMOUNT OF NOISE.

IN GENERAL, FOR ANY  $\underline{C}$  CHOOSE  $\underline{S}$  AS THE EIGENVECTOR OF  $\underline{C}$  WITH SMALLEST EIGENVALUE  $\lambda$ .

$$\underline{S} = \sqrt{E} \underline{V}_{\text{MIN}}, \text{ FOR } \underline{C} \underline{V}_{\text{MIN}} = \lambda_{\text{MIN}} \underline{V}_{\text{MIN}}.$$

SEE EX 4.5.

## MULTIPLE SIGNALS

### BINARY CASE

$$\begin{aligned} H_0: x(n) &= s_0(n) + w(n) & n=0, 1, \dots, N-1 \\ H_1: x(n) &= s_1(n) + w(n) & n=0, 1, \dots, N-1 \end{aligned}$$

$\uparrow$  WGN

$s_0(n), s_1(n)$  ARE DETERMINISTIC

AND KNOWN ( $s_0(n) = \cos 2\pi f_0 n$ ,

$$s_1(n) = \cos(2\pi f_0 n + \pi)$$

$\Rightarrow$  BPSK)