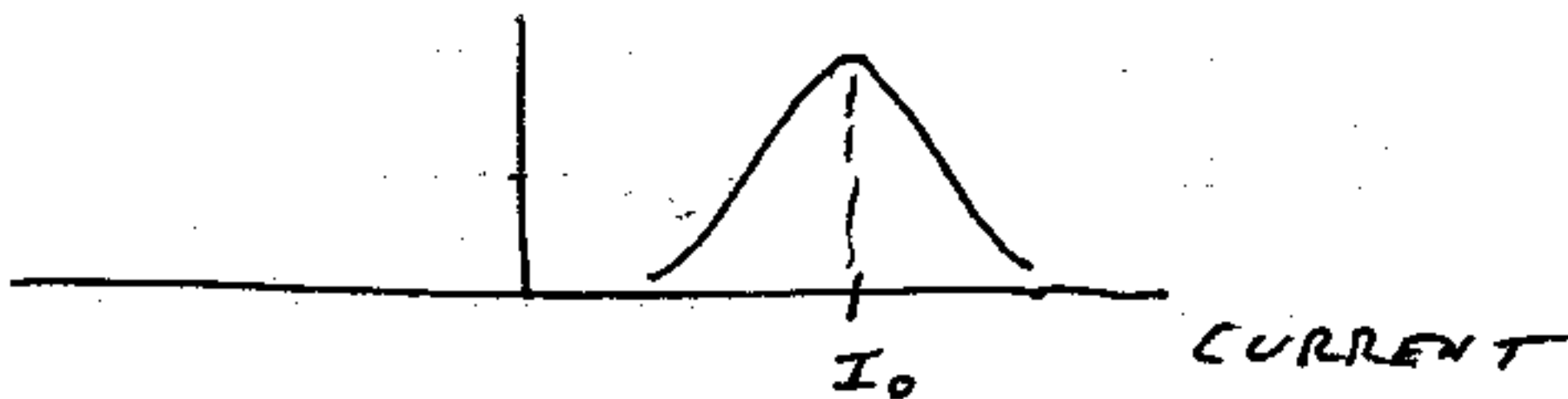


NON GAUSSIAN NOISE

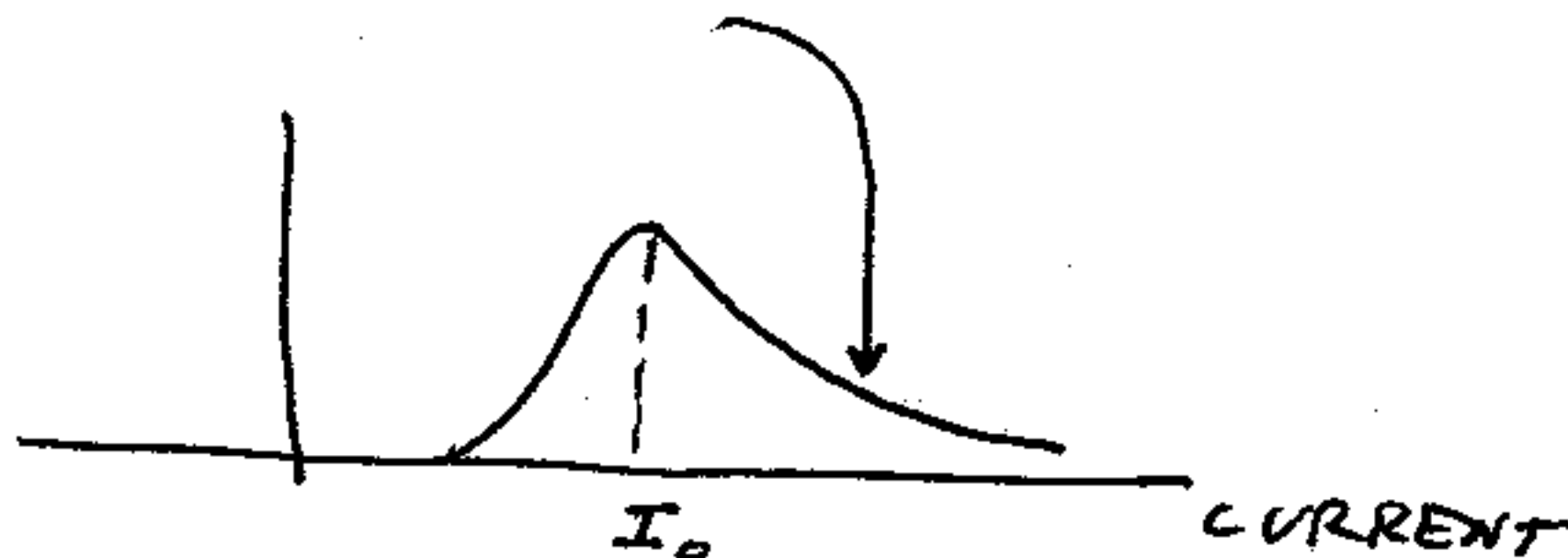
(CHAPTER 10)

GAUSSIAN PDF MODEL JUSTIFIED
BY CENTRAL LIMIT THEOREM

EXAMPLE : RESISTOR NOISE,
ELECTRONS ALL MOVING AT
ABOUT SAME VELOCITY \Rightarrow CURRENT
OR CHARGE PER UNIT TIME
IS SOME AVERAGE WITH
SMALL DEVIATIONS



WHAT HAPPENS IF WE HAVE SOME
" HOT RODDING " ELECTRONS?



NOISE IS NON GAUSSIAN.

OTHER EXAMPLES - HIGH LEVEL
BUT INFREQUENT EVENTS SUCH AS
ELECTROMAGNETIC NOISE SPIKES
DUE TO THUNDERSTORMS OR
ACOUSTIC SPIKES DUE TO ICEBERG
BREAKUP.

CONSIDER NOW DETERMINISTIC SIGNALS
IN NONGAUSSIAN NOISE.

NOISE CHARACTERISTICS

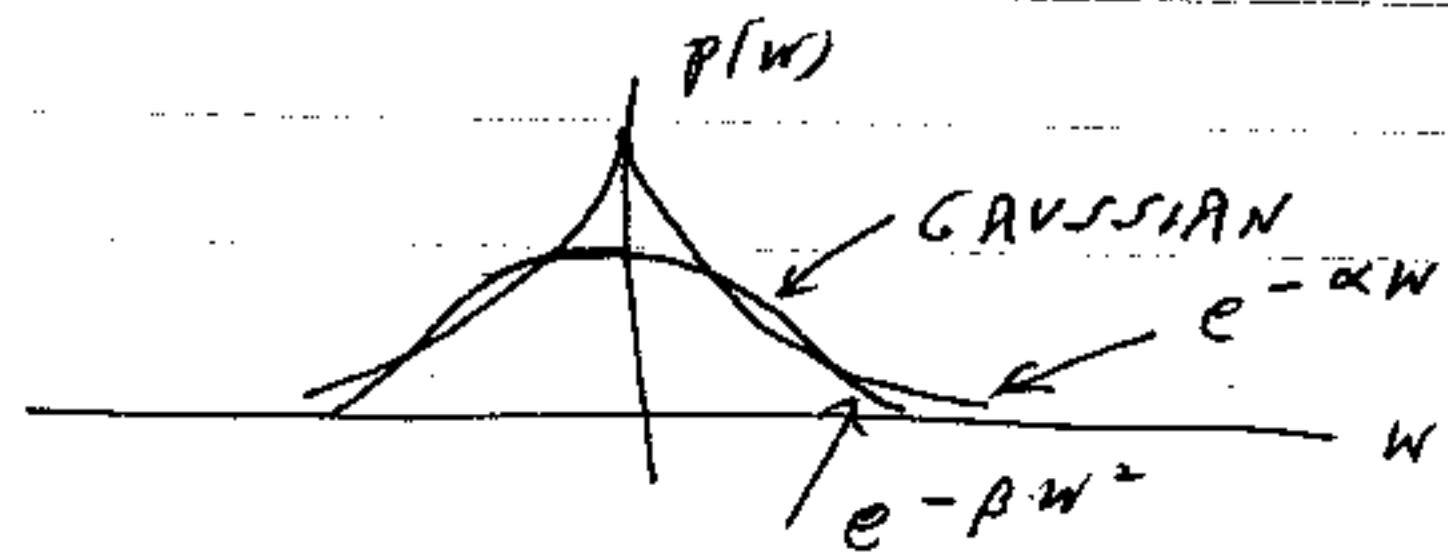
SIMPLEST MODEL IS IID

EXAMPLE : LAPLACIAN

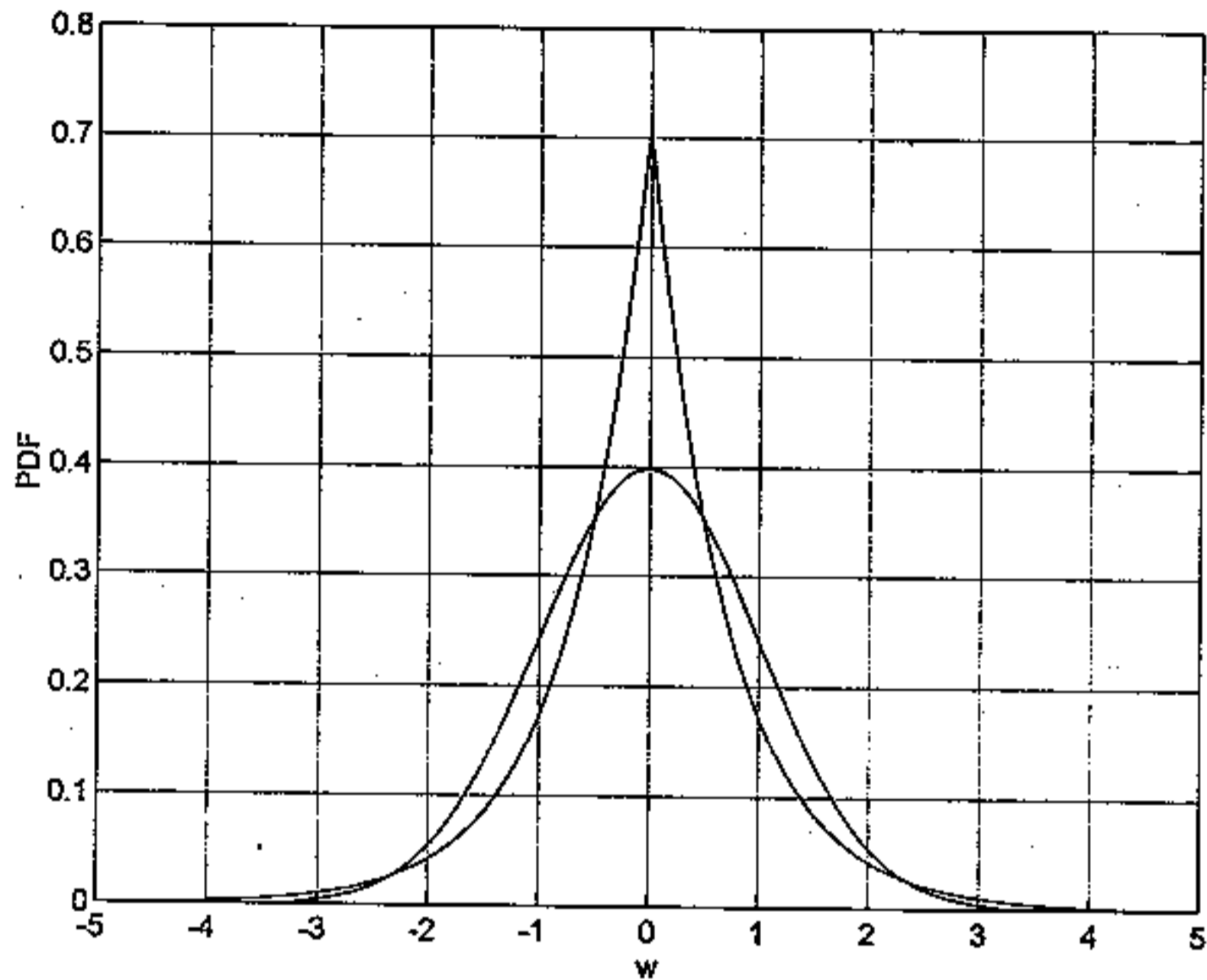
$$P(W|n) = \frac{1}{\sqrt{2\sigma^2}} e^{-\sqrt{\frac{2}{\sigma^2}} |W|n}$$

$n=0, 1, \dots, N-1$

ALL $W(n)$ HAVE SAME PDF AND ARE
INDEPENDENT.



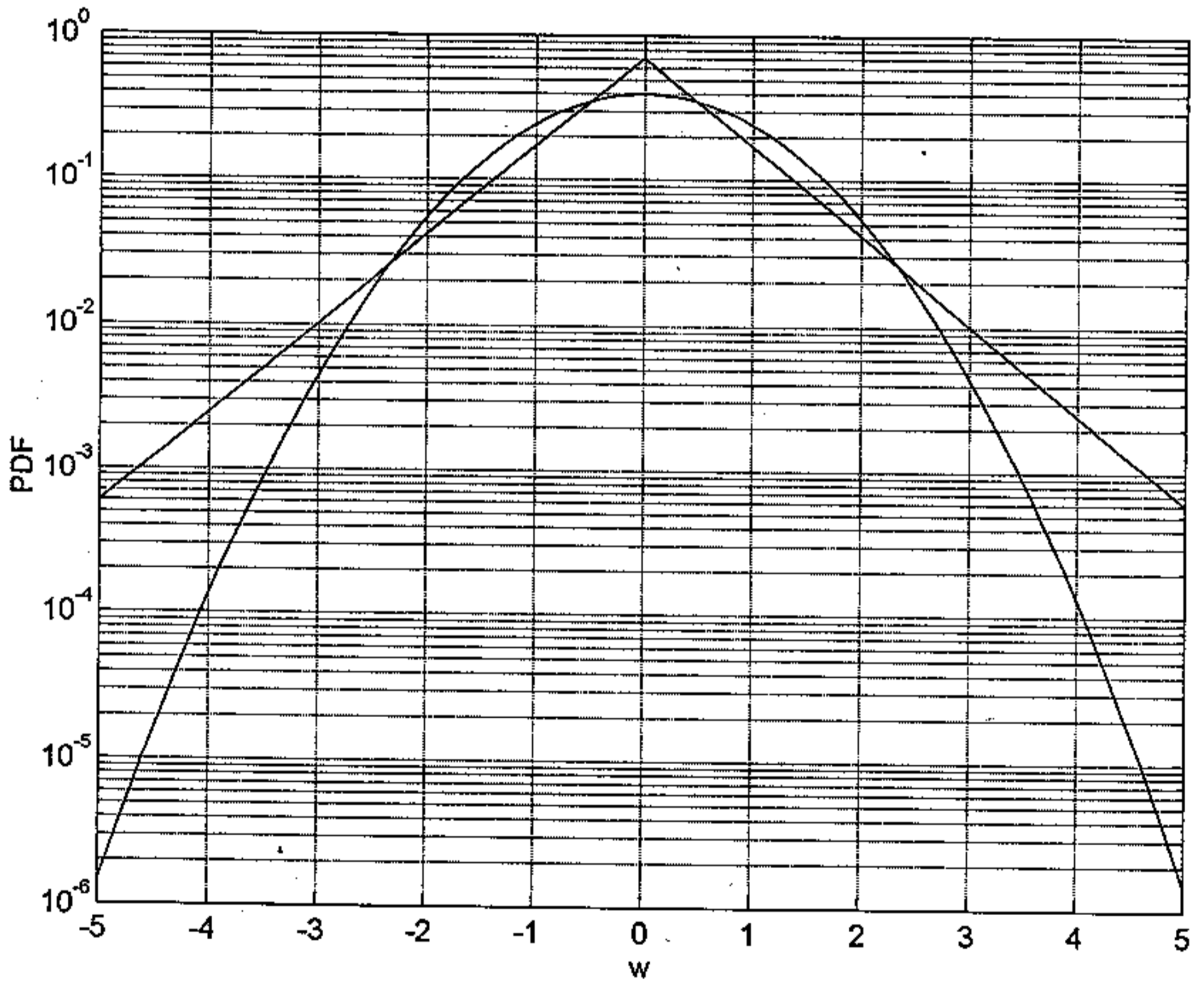
LAPLACIAN PDF "TAILS" ARE
HEAVIER \Rightarrow MORE HIGH LEVEL
EVENTS



a)

FIGURE 10.1 - GAUSSIAN VERSUS TYPICAL
NONGAUSSIAN PDF ($\sigma^2 = 1$)

NOTE: BOTH PDFS HAVE SAME VARIANCE
 σ^2 - OTHERWISE COMPARING
"APPLES AND ORANGES"



b)

FIGURE 10.1

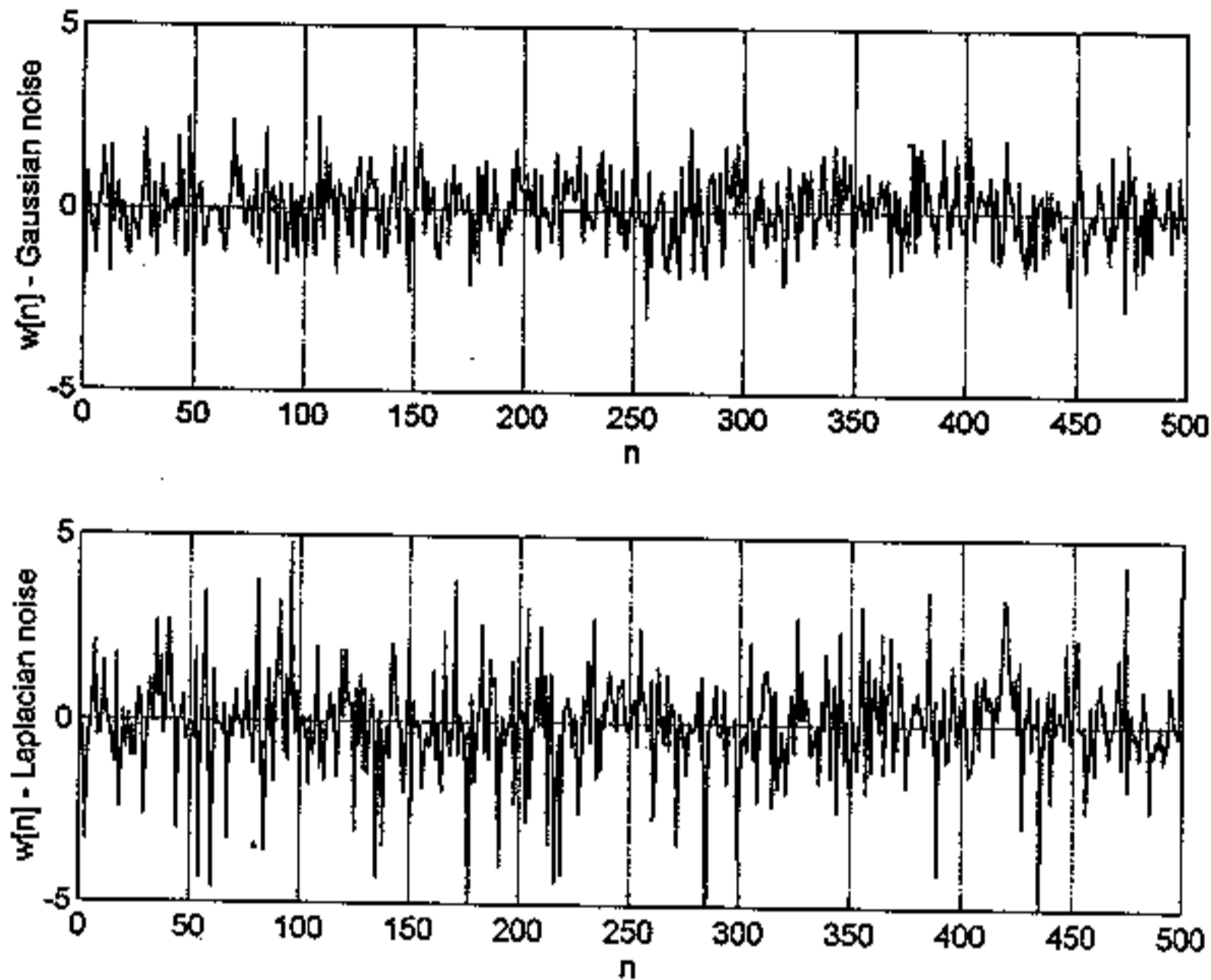


FIGURE 10.2 - REALIZATION OF GAUSSIAN AND NONGAUSSIAN NOISE PROCESSES ($\sigma^2 = 1$)

HIGH LEVEL EVENTS CALLED
SPIKES OR OUTLIERS - CAN
INCREASE PFA FOR GAUSSIAN
DESIGNED DETECTORS (NEED LIMITERS
OR CLIPPERS)

GENERALIZED GAUSSIAN PDFS

ENCOMPASSES LAPLACIAN, GAUSSIAN,
UNIFORM

$$p(w) = \frac{C_1(\beta)}{\sqrt{\sigma^2}} e^{-C_2(\beta) \left| \frac{w}{\sqrt{\sigma^2}} \right|^{\frac{2}{1+\beta}}} \quad -\infty < w < \infty$$

WHERE $C_1(\beta) = \frac{\Gamma^{\frac{1}{2}}\left(\frac{3}{2}(1+\beta)\right)}{(1+\beta)\Gamma^{\frac{3}{2}}\left(\frac{1}{2}(1+\beta)\right)}$

$$C_2(\beta) = \left[\frac{\Gamma\left(\frac{3}{2}(1+\beta)\right)}{\Gamma\left(\frac{1}{2}(1+\beta)\right)} \right]^{\frac{1}{1+\beta}}$$

$-1 < \beta \leq 1$ AND DETERMINES DEGREE OF
NONGAUSSIONITY

EXAMPLE:

- $\beta = 0 \Rightarrow$ GAUSSIAN
- $\beta = 1 \Rightarrow$ LAPLACIAN
- $\beta \rightarrow -1 \Rightarrow \approx$ UNIFORM

FOR $\beta > 0$ TAILS ARE HEAVIER

FOR $\beta < 0$ TAILS DIE OFF MORE
RAPIDLY

$$\beta = -\frac{1}{2} \Rightarrow p(w) \propto e^{-C_2(\beta) \left(\frac{w}{\sqrt{\sigma^2}} \right)^4}$$

SEE EXAMPLE 6.9

KNOWN DETERMINISTIC SIGNALS

EXAMPLE : DC LEVEL IN IID
NON GAUSSIAN NOISE

$$H_0: x(n) = w(n) \quad n = 0, 1, \dots, N-1$$

$$H_1: x(n) = A + w(n) \quad n = 0, 1, \dots, N-1$$

↑
KNOWN

$w(n)$ 'S ARE IID WITH KNOWN
PDF $p(w(n))$

NP DECIDES H_1 IF

$$L(x) = \frac{p(x; H_1)}{p(x; H_0)} > \gamma$$

DUE TO IID

$$L(x) = \frac{\prod_{n=0}^{N-1} p(x(n); H_1)}{\prod_{n=0}^{N-1} p(x(n); H_0)}$$

$$= \frac{\prod_{n=0}^{N-1} p(x[n] - A)}{\prod_{n=0}^{N-1} p(x[n])}$$

DECIDE H_1 IF

$$\ln L(x) = \sum_{n=0}^{N-1} \ln \frac{p(x[n] - A)}{p(x[n])} > \ln \gamma = \gamma'$$

$$\text{LET } g(x) = \ln \frac{p(x - A)}{p(x)}$$

DECIDE H_1 IF

$$\sum_{n=0}^{N-1} g(x[n]) > \gamma'$$

CLEARLY, $g(\cdot)$ DEPENDS ON NOISE PDF.

1) GAUSSIAN

$$g(x) = \ln \frac{\frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{1}{2\sigma^2}(x-A)^2}}{\frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{1}{2\sigma^2}x^2}}$$

$$= -\frac{1}{2\sigma^2} [(x-A)^2 - x^2] = -\frac{1}{2\sigma^2} (-2Ax + A^2)$$

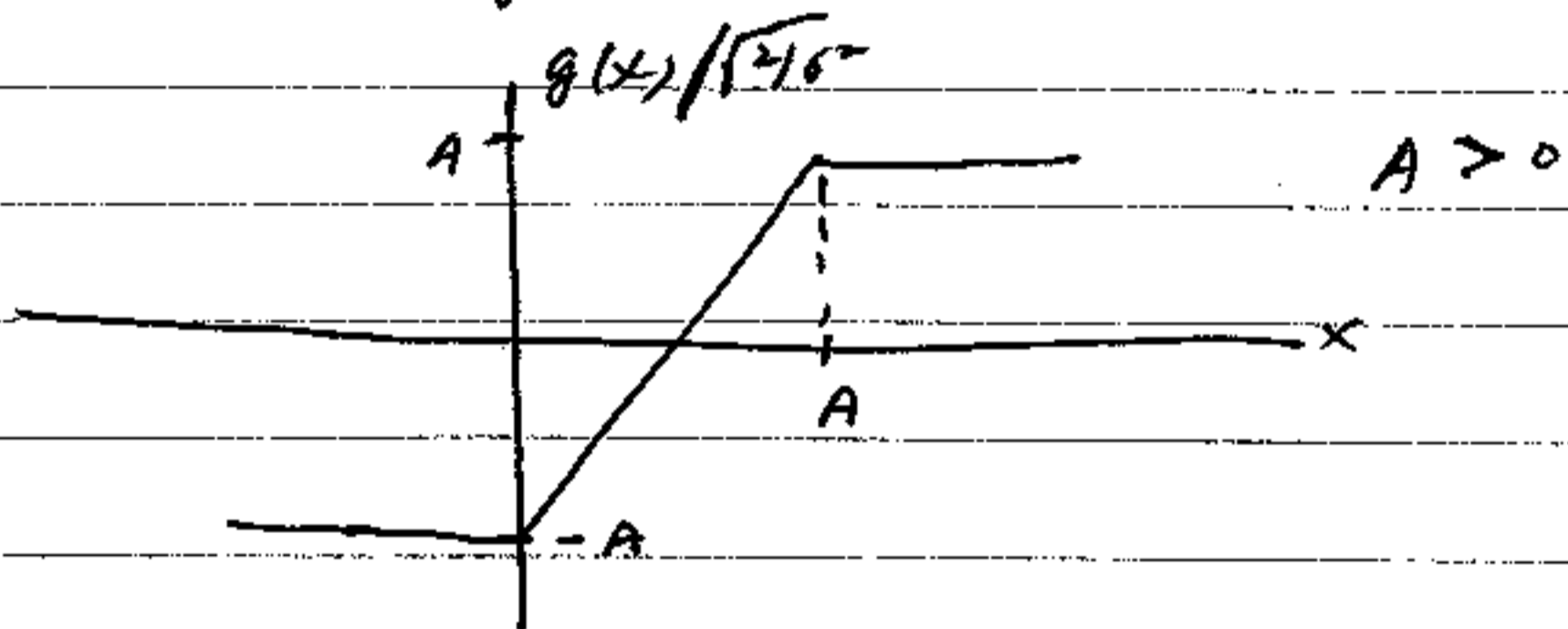
$$= \frac{A}{\sigma^2} x - \frac{A^2}{2\sigma^2}$$

$\Rightarrow g(\cdot)$ IS LINEAR (AFFINE) IN x
 STATISTIC BECOMES SAMPLE MEAN

2) LAPLACIAN

$$g(x) = \ln \frac{\frac{1}{\sqrt{2\sigma^2}} e^{-\sqrt{\frac{2}{\sigma^2}} |x-A|}}{\frac{1}{\sqrt{2\sigma^2}} e^{-\sqrt{\frac{2}{\sigma^2}} |x|}}$$

$$= \sqrt{\frac{2}{\sigma^2}} (|x| - |x-A|)$$



$g(\cdot)$ IS NOW NONLINEAR

A MORE INTUITIVE DETECTOR RESULTS
 FROM LETTING $y[n] = x[n] - A/2$

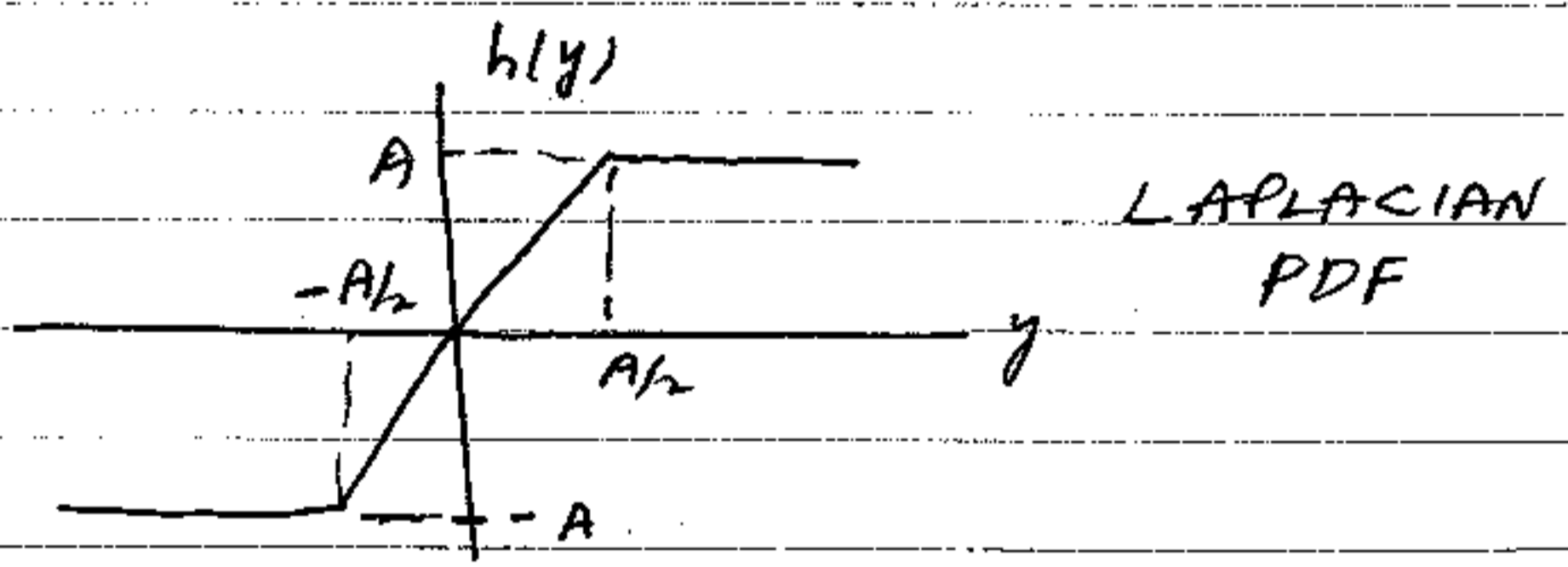
$$\left. \begin{aligned} E(y[n]; H_0) &= -A/2 \\ E(y[n]; H_1) &= A/2 \end{aligned} \right\} \text{SYMMETRIZED} \\ \text{VERSION}$$

$$\sum g(x|n) > \gamma'$$

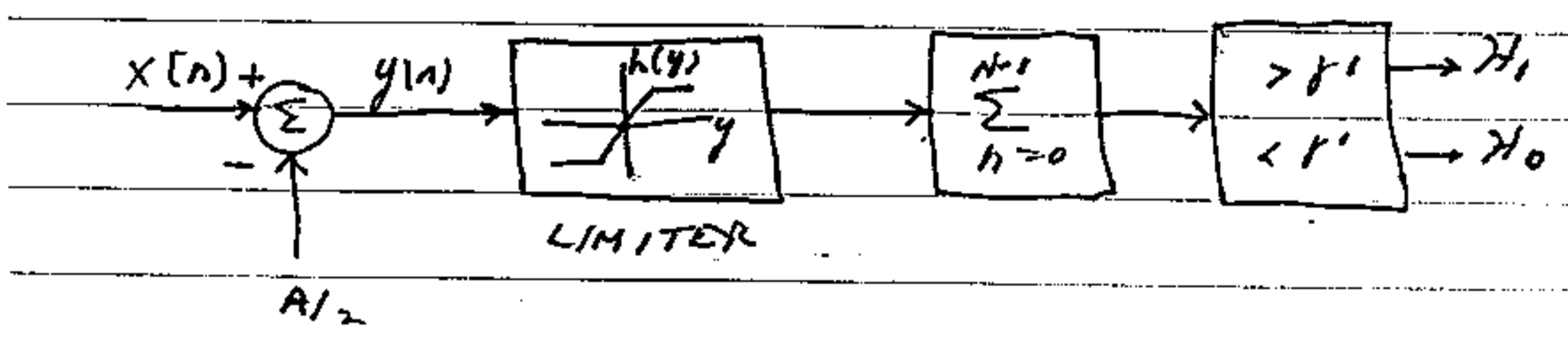
$$\Rightarrow \sum_{n=0}^{N-1} g(y|n) + A/2 > \gamma'$$

OR $\sum_{n=0}^{N-1} h(y|n) > \gamma'$

WHERE $h(y) = g(y + A/2)$
 $= \frac{\ln p(y - A/2)}{p(y + A/2)}$



DETECTOR FIRST SUBTRACTS $A/2$ FROM DATA, THEN CLIPS IT, AND THEN SUMS



CLIPPER ELIMINATES NOISE SPIKES

NOTE ALSO THAT CLIPPER IS
MEMORYLESS (INDEPENDENCE ASSUMPTION)
NONLINEAR TRANSFORMATION.

IN GENERAL FOR A SIGNAL $A \sin(n)$
WE DECIDE H_1 IF

$$\sum_{n=0}^{N-1} g_n(x[n]) > \gamma'$$

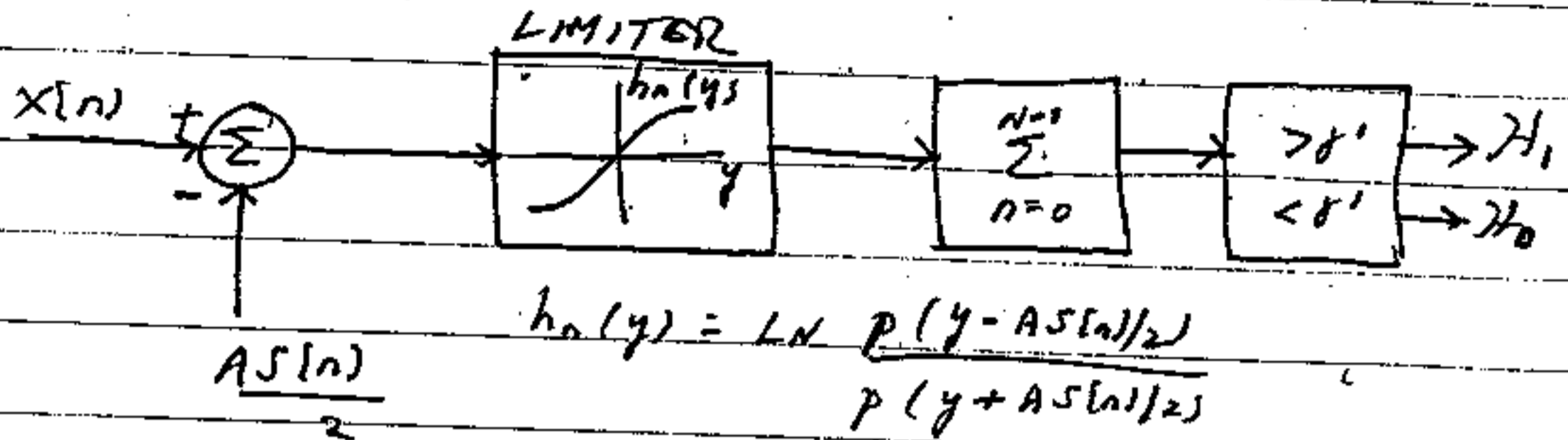
WHERE $g_n(x) = \ln \frac{p(x - A \sin(n))}{p(x)}$

LET $y[n] = x[n] - A \sin(n)/2$

DECIDE H_1 IF

$$\sum_{n=0}^{N-1} h_n(y[n]) > \gamma'$$

WHERE $h_n(y) = \ln \frac{p(y - A \sin(n)/2)}{p(y + A \sin(n)/2)}$



NOTE: IF $p(w)$ IS EVEN, $h_n(y)$ WILL BE ODD ($h_n(-y) = -h_n(y)$).

DETECTION PERFORMANCE

DIFFICULT DUE TO NONLINEARITY.

USE APPROXIMATION, VALID FOR WEAK SIGNALS ($A \rightarrow 0$ AND $N \rightarrow \infty$).

ASSUME SIGNAL IS $AS \sin$, WHERE $S \sin$ IS KNOWN AND $A > 0$. WHAT IS NP DETECTOR AND ITS PERFORMANCE AS $A \rightarrow 0$?

DECIDE H_1 IF $\sum_{n=0}^{N-1} g_n(x[n]) > \gamma'$

$$g_n(x) = \frac{\ln p(x - A s(n))}{p(x)}$$

AS $A \rightarrow 0$, NONLINEARITY CAN BE
LINEARIZED ABOUT $A=0$

$$g_n(x) \approx g_n(x) \Big|_{A=0} + \frac{\partial g_n(x)}{\partial A} \Big|_{A=0} A$$

$$= 0 + \frac{\frac{d p(x - A s(n))}{d A}}{p(x - A s(n))} \Big|_{A=0} A$$

$$= \frac{\frac{d p(w)}{d w} (-s(n))}{p(w)} \Big|_{\substack{w=x-A s(n), \\ A=0}} A$$

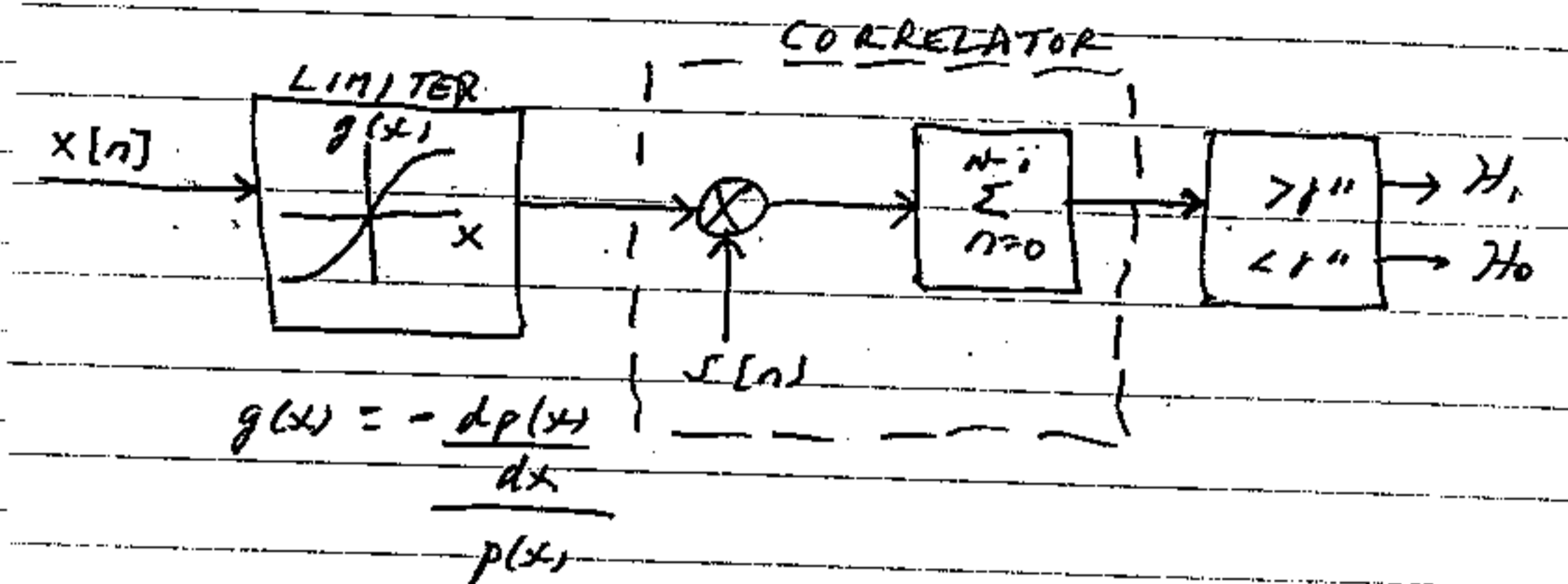
$$= \frac{\frac{d p(x)}{d x} s(n) A}{p(x)}$$

DECIDE H_1 IF

$$\sum_{n=0}^{N-1} g_n(x(n)) \approx \sum_{n=0}^{N-1} - \frac{\frac{d p(x(n))}{d x(n)} s(n) A}{p(x(n))} > \gamma'$$

OR SINCE $A > 0$

$$T(x) = \sum_{n=0}^{N-1} - \frac{\frac{d p(x(n))}{d x(n)} s(n)}{p(x(n))} > \gamma''$$



WHAT DO WE GET FOR GAUSSIAN NOISE?

NON-LINEARITY IS NOW

$$g(x) = - \frac{dp(x)}{dx} \cdot \frac{1}{p(x)} = - \frac{d \ln p(x)}{dx}$$

THIS DETECTOR IS THE SMALL SIGNAL NP DETECTOR \Rightarrow CALLED LOCALLY OPTIMUM DETECTOR. IF A IS LARGE, WILL NOT BE OPTIMUM.

CAN NOW FIND PERFORMANCE
 (ASYMPTOTIC NP DETECTOR AS
 $A \rightarrow 0, N \rightarrow \infty$)

SEE APPENDIX 10A ($T(x)$ STILL
 NONLINEAR IN DATA BUT CAN USE
 CENTRAL LIMIT THEOREM SINCE

$$\frac{\frac{d p(x|n)}{d x|n}}{p(x|n)} \quad \begin{array}{l} \text{IS IID UNDER } H_0 \\ x|n = w|n \\ \approx \text{IID UNDER } H_1 \\ x|n = A s|n + w|n \end{array}$$

RESULT:

$$T(x) = \sum_{n=0}^{N-1} \frac{\frac{d p(x|n)}{d x|n}}{p(x|n)} s|n = \sum_{n=0}^{N-1} g(x|n) s|n$$

$$\stackrel{a}{\approx} N\left(0, \dot{c}(A) \sum_{n=0}^{N-1} s^2|n\right) H_0$$

$$N\left(A \dot{c}(A) \sum_{n=0}^{N-1} s^2|n, \dot{c}(A) \sum_{n=0}^{N-1} s^2|n\right) H_1$$

$$\text{WHERE } \dot{c}(A) = \int_{-\infty}^{\infty} \frac{\left(\frac{d p(w)}{d w}\right)^2}{p(w)} dw$$

$i(A)$ = FISHER INFORMATION FOR
A BASED ON SINGLE SAMPLE.

THIS IS JUST MEAN-SHIFTED GAUSS -
GAUSS PROBLEM

$$\Rightarrow P_D = Q(Q^{-1}(P_{FA}) - \sqrt{d^2})$$

WHERE

$$d^2 = \frac{(E(T; \mathcal{H}_1) - E(T; \mathcal{H}_0))^2}{\text{VAR}(T; \mathcal{H}_0)}$$

$$= A^2 i(A) \sum_{n=0}^{N-1} \sigma^2(n)$$

EXAMPLE : GAUSSIAN PDF, IID
SAMPLES

DECIDE \mathcal{H}_1 IF

$$T(x) = \sum_{n=0}^{N-1} g(x|n) \sigma(n) > \gamma'$$

$$g(x) = \frac{-\frac{d p(x)}{dx}}{p(x)} = -\frac{d \ln p(x)}{dx}$$

$$p(x) = \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{1}{2\sigma^2}x^2}$$

$$g(x) = -\frac{d}{dx} \left(-\frac{1}{2\sigma^2}x^2 \right)$$

$$= x/\sigma^2$$

$$T(x) = 1/\sigma^2 \sum_{n=0}^{N-1} x(n) s(n)$$

USUAL CORRELATOR

(FOR THIS CASE NP \equiv WEAK SIGNAL NP)

$$d^2 = A^2 i(A) \sum_{n=0}^{N-1} s^2(n)$$

$$i(A) = \int \frac{\left(\frac{dp(w)}{dw} \right)^2}{p(w)} dw$$

$$= \int \left(\frac{d \ln p(w)}{dw} \right)^2 p(w) dw$$

$$= E \left[\left(\frac{d \ln p(w)}{dw} \right)^2 \right]$$

$$\begin{aligned}
 &= E [g^2(w)] \\
 &= E (w^2 / \sigma^4) = 1 / \sigma^2 \\
 d^2 &= \frac{A^2 \sum s^2(n)}{\sigma^2} = E / \sigma^2
 \end{aligned}$$

AS EXPECTED

ASIDE: $i(A)$ CALLED INTRINSIC ACCURACY OF PDF, IS THE FISHER INFORMATION FOR THE PARAMETER μ IF $p(x; \mu) = p_w(x - \mu)$

GAUSSIAN PDF HAS SMALLEST $i(A)$ FOR ALL PDFS WITH SAME VARIANCE \Rightarrow GAUSSIAN PDF IS WORST CASE (SMALLEST d^2)

EXAMPLE: NONGAUSSIAN NOISE

CONSIDER LAPLACIAN NOISE AND A DC LEVEL $A > 0$

DECIDE H_1 IF

$$T(\underline{x}) = \sum_{n=0}^{N-1} \frac{d p(x|n)}{dx|n} \frac{A > \gamma'}{p(x|n)}$$

OR

$$T(\underline{x}) = \sum_{n=0}^{N-1} \frac{d p(x|n)}{dx|n} > \gamma'' \frac{p(x|n)}{p(x|n)}$$

BUT

$$p(w) = \frac{1}{\sqrt{2\sigma^2}} e^{-\sqrt{2/\sigma^2} |w|}$$

$$g(x) = \frac{-dp(x)/dx}{p(x)} = -\frac{d \ln p(x)}{dx}$$

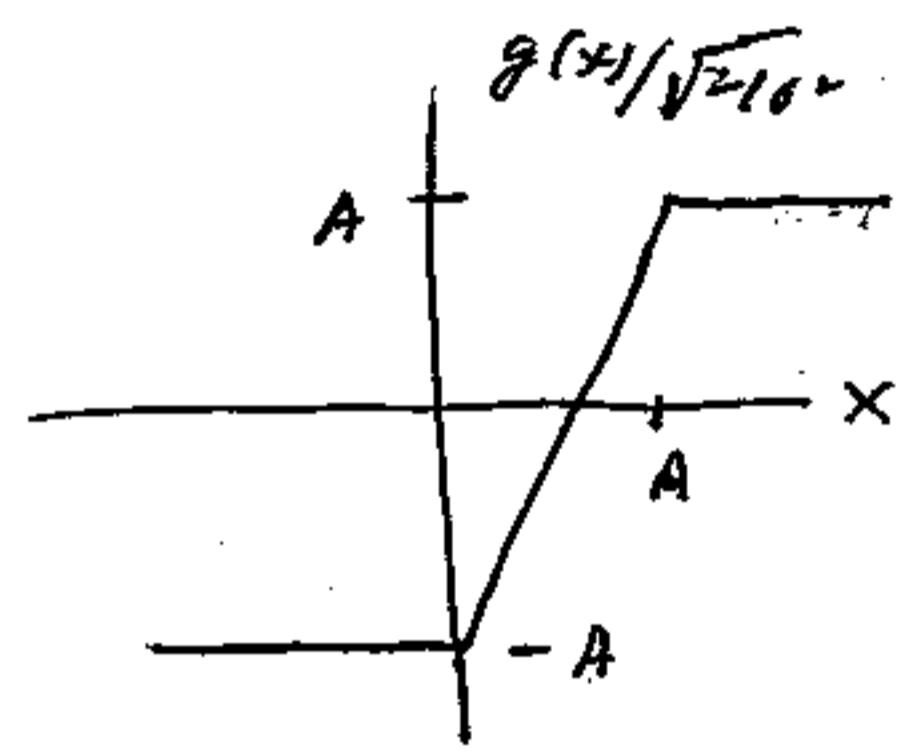
$$= -\frac{d}{dx} (-\sqrt{2/\sigma^2} |x|)$$

$$= \sqrt{2/\sigma^2} \frac{d|x|}{dx} = \sqrt{2/\sigma^2} \text{sgn}(x)$$

$$T(\underline{x}) = \sqrt{2/\sigma^2} \sum_{n=0}^{N-1} \text{sgn}(x|n)$$

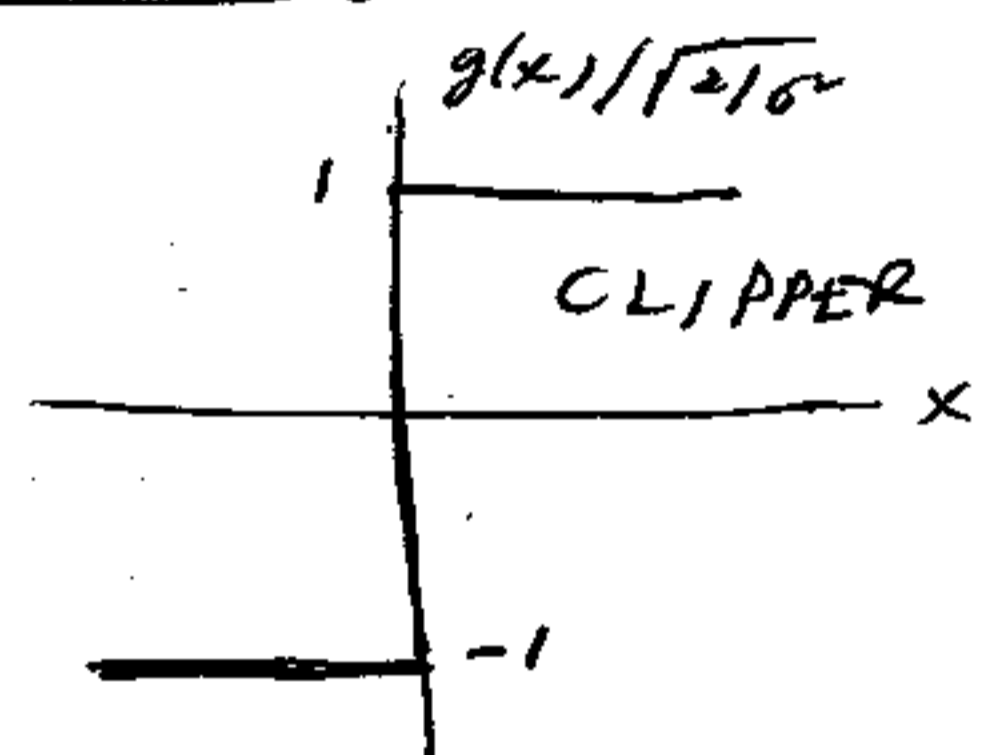
WEAK SIGNAL DETECTOR FOR DC LEVEL ($A > 0$) IN IID LAPLACIAN NOISE: ADD SAMPLE SIGNS.

CALLED A SIGN DETECTOR.



A KNOWN - NP

$A \rightarrow 0$
→



KNOWN OR UNKNOWN, $A > 0$
WEAK SIGNAL NP

PERFORMANCE (AS $A \rightarrow 0$) IS

$$P_D = Q(Q^{-1}(P_{FA}) - \sqrt{d^2})$$

$$d^2 = A^2 \dot{c}(A) \sum \sigma^2(n)$$

$$= A^2 \dot{c}(A) N$$

TO FIND $\dot{c}(A)$:

$$\begin{aligned}
 i(A) &= E(g^2(w)) \\
 &= E((\sqrt{2}/\sigma - s \cos(w))^2) \\
 &= 2/\sigma^2 E(1) = 2/\sigma^2 \\
 &= \text{TWICE THAT FOR GAUSSIAN NOISE}
 \end{aligned}$$

$$d_{LAP}^2 = \frac{2NA^2}{\sigma^2} = 2 d_{GAUSSIAN}^2$$

DETERMINISTIC SIGNALS WITH UNKNOWN PARAMETERS

RAO TEST IS EASIER TO IMPLEMENT THAN GLRT.

CONSIDER

$$\begin{aligned}
 H_0: & \quad x[n] = w[n] & n = 0, 1, \dots, N-1 \\
 H_1: & \quad x[n] = A s[n] + w[n] \\
 & \quad \quad \quad \uparrow & \quad \quad \quad \uparrow \\
 & \quad \quad \quad \text{UNKNOWN} & \quad \quad \quad \text{IID NONGAUSSIAN} \\
 & \quad \quad \quad -\infty < A < \infty & \quad \quad \quad p(w[n])
 \end{aligned}$$